Lecture - 10
Image Enhancement in the Frequency Domain

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Background

- Any function that **periodically** repeats itself can be expressed as the **sum** of sines and/or cosines of different frequencies, each multiplied by a different coefficient (**Fourier series**).

- Even functions that are **not periodic** (but whose area under the curve is finite) can be expressed as the **integral** of sines and/or cosines multiplied by a weighting function (**Fourier transform**).
Background

- The **frequency domain** refers to the plane of the two dimensional discrete Fourier transform of an image.
- The purpose of the Fourier transform is to represent a signal as a linear combination of sinusoidal signals of various frequencies.

**Figure 4.1** The function at the bottom is the sum of the four functions above it. Fourier’s idea in 1807 that periodic functions could be represented as a weighted sum of sines and cosines was met with skepticism.
Introduction to the Fourier Transform and the Frequency Domain

- The one-dimensional Fourier transform and its inverse
  - Fourier transform (continuous case)
    \[
    F(u) = \int_{-\infty}^{\infty} f(x)e^{-j2\pi u x} \, dx \quad \text{where} \quad j = \sqrt{-1}
    \]
  - Inverse Fourier transform:
    \[
    f(x) = \int_{-\infty}^{\infty} F(u)e^{j2\pi u x} \, du
    \]
- The two-dimensional Fourier transform and its inverse
  - Fourier transform (continuous case)
    \[
    F(u, v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y)e^{-j2\pi (ux+vy)} \, dx \, dy
    \]
  - Inverse Fourier transform:
    \[
    f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v)e^{j2\pi (ux+vy)} \, du \, dv
    \]

\[e^{j\theta} = \cos \theta + j \sin \theta\]
Introduction to the Fourier Transform and the Frequency Domain

- The one-dimensional Fourier transform and its inverse
  - Fourier transform (discrete case) DCT
    \[ F(u) = \frac{1}{M} \sum_{x=0}^{M-1} f(x)e^{-j \frac{2\pi ux}{M}} \quad \text{for } u = 0,1,2,\ldots, M - 1 \]
  - Inverse Fourier transform:
    \[ f(x) = \sum_{u=0}^{M-1} F(u)e^{j \frac{2\pi ux}{M}} \quad \text{for } x = 0,1,2,\ldots, M - 1 \]
Since  \( e^{j\theta} = \cos \theta + j \sin \theta \) and the fact \( \cos(-\theta) = \cos \theta \)

then discrete Fourier transform can be redefined

\[
F(u) = \frac{1}{M} \sum_{x=0}^{M-1} f(x)[\cos 2\pi ux / M - j \sin 2\pi ux / M]
\]

for \( u = 0,1,2,...,M-1 \)

- Frequency (time) domain: the domain (values of \( u \)) over which the values of \( F(u) \) range; because \( u \) determines the frequency of the components of the transform.
- Frequency (time) component: each of the \( M \) terms of \( F(u) \).
• \( F(u) \) can be expressed in polar coordinates:

\[
F(u) = |F(u)| e^{i \phi(u)}
\]

where \( |F(u)| = \left[ R^2(u) + I^2(u) \right]^{1/2} \) (magnitude or spectrum)

\[
\phi(u) = \tan^{-1} \left[ \frac{I(u)}{R(u)} \right] \) (phase angle or phase spectrum)

- \( R(u) \): the real part of \( F(u) \)
- \( I(u) \): the imaginary part of \( F(u) \)

• Power spectrum:

\[
P(u) = |F(u)|^2 = R^2(u) + I^2(u)
\]

Also referred to spectral density
The One-Dimensional Fourier Transform Example

**Image Processing**

**FIGURE 4.2** (a) A discrete function of $M$ points, and (b) its Fourier spectrum. (c) A discrete function with twice the number of nonzero points, and (d) its Fourier spectrum.
The One-Dimensional Fourier Transform
Some Examples

• The transform of a constant function is a DC value only.

• The transform of a delta function is a constant.
The One-Dimensional Fourier Transform
Some Examples

• The transform of an infinite train of delta functions spaced by T is an infinite train of delta functions spaced by 1/T.

• The transform of a cosine function is a positive delta at the appropriate positive and negative frequency.
The One-Dimensional Fourier Transform
Some Examples

• The transform of a sin function is a negative complex delta function at the appropriate positive frequency and a negative complex delta at the appropriate negative frequency.

• The transform of a square pulse is a sinc function.
Introduction to the Fourier Transform and the Frequency Domain

- The two-dimensional Fourier transform and its inverse
  - Fourier transform (discrete case) DFT
    \[ F(u, v) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y)e^{-j2\pi(ux/M + vy/N)} \]
    for \( u = 0,1,2,\ldots, M - 1, v = 0,1,2,\ldots, N - 1 \)
  - Inverse Fourier transform:
    \[ f(x, y) = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F(u, v)e^{j2\pi(ux/M + vy/N)} \]
    for \( x = 0,1,2,\ldots, M - 1, y = 0,1,2,\ldots, N - 1 \)

- \( u, v \) : the transform or frequency variables
- \( x, y \) : the spatial or image variables
Introduction to the Fourier Transform and the Frequency Domain

- We define the Fourier spectrum, phase angle, and power spectrum as follows:

\[
|F(u, v)| = \left[ R^2(u, v) + I^2(u, v) \right]^{1/2} \quad \text{(spectrum)}
\]

\[
\phi(u, v) = \tan^{-1} \left( \frac{I(u, v)}{R(u, v)} \right) \quad \text{(phase angle)}
\]

\[
P(u, v) = |F(u, v)|^2 = R^2(u, v) + I^2(u, v) \quad \text{(power spectrum)}
\]

- \( R(u, v) \): the real part of \( F(u, v) \)
- \( I(u, v) \): the imaginary part of \( F(u, v) \)
Some properties of Fourier transform:

\[ \mathcal{F}[f(x, y)(-1)^{x+y}] = F(u - \frac{M}{2}, v - \frac{N}{2}) \] (shift)

\[ F(0,0) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \] (average)

\[ F(u, v) = F^*(-u, -v) \] (conjugate symmetric)

\[ |F(u, v)| = |F(-u, -v)| \] (symmetric)
The 2D DFT $F(u,v)$ can be obtained by
1. taking the 1D DFT of every row of image $f(x,y)$, $F(u,y)$,
2. taking the 1D DFT of every column of $F(u,y)$
The Two-Dimensional DFT and Its Inverse

**FIGURE 4.3**
(a) Image of a $20 \times 40$ white rectangle on a black background of size $512 \times 512$ pixels.
(b) Centered Fourier spectrum shown after application of the log transformation given in Eq. (3.2-2). Compare with Fig. 4.2.
The Two-Dimensional DFT and Its Inverse
The Property of Two-Dimensional DFT Rotation
The Property of Two-Dimensional DFT Linear Combination

\[ 0.25 \times A + 0.75 \times B \]
Expanding the original image A by a factor of n (n=2), filling the empty new values with zeros (B), results in the same DFT.
Two-Dimensional DFT with Different Functions

Sine wave

Rectangle

Its DFT

Its DFT
Two-Dimensional DFT with Different Functions

2D Gaussian function

Its DFT

Impulses

Its DFT
Filtering in the Frequency Domain

FIGURE 4.4
(a) SEM image of a damaged integrated circuit.
(b) Fourier spectrum of (a).
(Original image courtesy of Dr. J. M. Hudak, Brockhouse Institute for Materials Research, McMaster University, Hamilton, Ontario, Canada.)
Basics of Filtering in the Frequency Domain

Figure 4.5 Basic steps for filtering in the frequency domain.
Basics of Filtering in the Frequency Domain

1. Multiply the input image by \((-1)^{x+y}\) to center the transform, as indicated in Eq. (4.2-21).
2. Compute \(F(u, v)\), the DFT of the image from (1).
3. Multiply \(F(u, v)\) by a filter function \(H(u, v)\).
4. Compute the inverse DFT of the result in (3).
5. Obtain the real part of the result in (4).
6. Multiply the result in (5) by \((-1)^{x+y}\).
Some Basic Filters and Their Functions

- Multiply all values of $F(u,v)$ by the filter function (notch filter):

$$H(u,v) = \begin{cases} 
0 & \text{if } (u,v) = (M/2, N/2) \\
1 & \text{otherwise.}
\end{cases}$$

- All this filter would do is set $F(0,0)$ to zero (force the average value of an image to zero) and leave all frequency components of the Fourier transform untouched.

FIGURE 4.6
Result of filtering the image in Fig. 4.4(a) with a notch filter that set to 0 the $F(0, 0)$ term in the Fourier transform.
Some Basic Filters and Their Functions

Lowpass filter

Highpass filter

FIGURE 4.7 (a) A two-dimensional lowpass filter function. (b) Result of lowpass filtering the image in Fig. 4.4(a). (c) A two-dimensional highpass filter function. (d) Result of highpass filtering the image in Fig. 4.4(a).
Some Basic Filters and Their Functions

**FIGURE 4.8**
Result of highpass filtering the image in Fig. 4.4(a) with the filter in Fig. 4.7(c), modified by adding a constant of one-half the filter height to the filter function. Compare with Fig. 4.4(a).
Correspondence between Filtering in the Spatial and Frequency Domain

- **Convolution theorem:**
  - The discrete convolution of two functions $f(x,y)$ and $h(x,y)$ of size $M \times N$ is defined as
    \[
    f(x, y) \ast h(x, y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f(m, n)h(x - m, y - n)
    \]
  - Let $F(u,v)$ and $H(u,v)$ denote the Fourier transforms of $f(x,y)$ and $h(x,y)$, then
    \[
    f(x, y) \ast h(x, y) \Leftrightarrow F(u, v)H(u, v) \quad \text{Eq. (4.2-31)}
    \]
    \[
    f(x, y)h(x, y) \Leftrightarrow F(u, v) \ast H(u, v) \quad \text{Eq. (4.2-32)}
    \]
Correspondence between Filtering in
the Spatial and Frequency Domain

- \( A\delta(x-x_0, y-y_0) \) : an impulse function of strength \( A \), located at coordinates \((x_0, y_0)\)

\[
\sum_{x=0}^{M-1} \sum_{y=0}^{N-1} s(x, y) A\delta(x-x_0, y-y_0) = As(x_0, y_0)
\]

where \( \delta(x, y) \) : a unit impulse located at the origin

- The Fourier transform of a unit impulse at the origin (Eq. (4.2-35)):

\[
F(u, v) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} \delta(x, y) e^{-j2\pi(ux/M+vy/N)} = \frac{1}{MN}
\]
Correspondence between Filtering in the Spatial and Frequency Domain

- Let \( f(x, y) = \delta(x, y) \), then the convolution (Eq. (4.2-36))

\[
f(x, y) * h(x, y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} \delta(m, n)h(x - m, y - n)
\]

\[
= \frac{1}{MN} h(x, y)
\]

- Combine Eqs. (4.2-35) (4.2-36) with Eq. (4.2-31), we obtain

\[
f(x, y) * h(x, y) \Leftrightarrow F(u, v)H(u, v)
\]

\[
\delta(x, y) * h(x, y) \Leftrightarrow \mathcal{F}^{-1}\{\mathcal{F}\{\delta(x, y)\}H(u, v)\}
\]

\[
\frac{1}{MN} h(x, y) \Leftrightarrow H(u, v)
\]

\[
\frac{1}{MN} H(u, v)
\]
Correspondence between Filtering in the Spatial and Frequency Domain

• Let $H(u)$ denote a frequency domain, Gaussian filter function given the equation

$$H(u) = Ae^{-u^2/2\sigma^2}$$

where $\sigma$: the standard deviation of the Gaussian curve.

• The corresponding filter in the spatial domain is

$$h(x) = \sqrt{2\pi \sigma} Ae^{-2\pi^2 \sigma^2 x^2}$$

• Note: Both the forward and inverse Fourier transforms of a Gaussian function are real Gaussian functions.
Correspondence between Filtering in the Spatial and Frequency Domain

**FIGURE 4.9**
(a) Gaussian frequency domain lowpass filter.
(b) Gaussian frequency domain highpass filter.
(c) Corresponding lowpass spatial filter.
(d) Corresponding highpass spatial filter. The masks shown are used in Chapter 3 for lowpass and highpass filtering.
Correspondence between Filtering in the Spatial and Frequency Domain

• One very useful property of the Gaussian function is that both it and its Fourier transform are real valued; there are no complex values associated with them.
• In addition, the values are always positive. So, if we convolve an image with a Gaussian function, there will never be any negative output values to deal with.
• There is also an important relationship between the widths of a Gaussian function and its Fourier transform. If we make the width of the function smaller, the width of the Fourier transform gets larger. This is controlled by the variance parameter $\sigma^2$ in the equations.
• These properties make the Gaussian filter very useful for lowpass filtering an image. The amount of blur is controlled by $\sigma^2$. It can be implemented in either the spatial or frequency domain.
• Other filters besides lowpass can also be implemented by using two different sized Gaussian functions.
Smoothing Frequency-Domain Filters

- The basic model for filtering in the frequency domain
  \[ G(u, v) = H(u, v)F(u, v) \]
  where \( F(u,v) \): the Fourier transform of the image to be smoothed
  \( H(u,v) \): a filter transfer function

- Smoothing is fundamentally a lowpass operation in the frequency domain.
- There are several standard forms of lowpass filters (LPF).
  - Ideal lowpass filter
  - Butterworth lowpass filter
  - Gaussian lowpass filter
Ideal Lowpass Filters (ILPFs)

- The simplest lowpass filter is a filter that “cuts off” all high-frequency components of the Fourier transform that are at a distance greater than a specified distance $D_0$ from the origin of the transform.

- The transfer function of an ideal lowpass filter

$$H(u,v) = \begin{cases} 
1 & \text{if } D(u,v) \leq D_0 \\
0 & \text{if } D(u,v) > D_0 
\end{cases}$$

where $D(u,v)$: the distance from point $(u,v)$ to the center of their frequency rectangle

$$D(u,v) = \left[ \left( u - M / 2 \right)^2 + \left( v - N / 2 \right)^2 \right]^{\frac{1}{2}}$$
Ideal Lowpass Filters (ILPFs)

**Figure 4.10** (a) Perspective plot of an ideal lowpass filter transfer function. (b) Filter displayed as an image. (c) Filter radial cross section.
Ideal Lowpass Filters (ILPFs)

**FIGURE 4.11** (a) An image of size $500 \times 500$ pixels and (b) its Fourier spectrum. The superimposed circles have radii values of 5, 15, 30, 80, and 230, which enclose 92.0, 94.6, 96.4, 98.0, and 99.5% of the image power, respectively.
**Ideal Lowpass Filters**

**FIGURE 4.12** (a) Original image. (b)–(f) Results of ideal lowpass filtering with cutoff frequencies set at radii values of 5, 15, 30, 80, and 230, as shown in Fig. 4.11(b). The power removed by these filters was 8, 5.4, 3.6, 2, and 0.5% of the total, respectively.
Butterworth Lowpass Filters (BLPFs)  
With order $n$

$$H(u, v) = \frac{1}{1 + \left[ \frac{D(u, v)}{D_0} \right]^{2n}}$$

**FIGURE 4.14** (a) Perspective plot of a Butterworth lowpass filter transfer function. (b) Filter displayed as an image. (c) Filter radial cross sections of orders 1 through 4.
Butterworth Lowpass Filters (BLPFs)

\[ n = 2 \]

\[ D_0 = 5, 15, 30, 80, \text{and } 230 \]

**FIGURE 4.15** (a) Original image. (b)–(f) Results of filtering with BLPFs of order 2, with cutoff frequencies at radii of 5, 15, 30, 80, and 230, as shown in Fig. 4.11(b). Compare with Fig. 4.12.
Butterworth Lowpass Filters (BLPFs)

Spatial Representation

**FIGURE 4.16** (a)–(d) Spatial representation of BLPFs of order 1, 2, 5, and 20, and corresponding gray-level profiles through the center of the filters (all filters have a cutoff frequency of 5). Note that ringing increases as a function of filter order.
Gaussian Lowpass Filters (FLPFs)

\[ H(u, v) = e^{-D^2(u,v)/2D_0^2} \]

**FIGURE 4.17** (a) Perspective plot of a GLPF transfer function. (b) Filter displayed as an image. (c) Filter radial cross sections for various values of \( D_0 \).
Gaussian Lowpass Filters (FLPFs)

\[ D_0 = 5, 15, 30, 80, \text{and } 230 \]

**FIGURE 4.18** (a) Original image. (b)–(f) Results of filtering with Gaussian lowpass filters with cutoff frequencies set at radii values of 5, 15, 30, 80, and 230, as shown in Fig. 4.11(b). Compare with Figs. 4.12 and 4.15.
**FIGURE 4.19**
(a) Sample text of poor resolution (note broken characters in magnified view).
(b) Result of filtering with a GLPF (broken character segments were joined).

Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.
**Additional Examples of Lowpass Filtering**

**Figure 4.20** (a) Original image (1028 × 732 pixels). (b) Result of filtering with a GLPF with \( D_0 = 100 \). (c) Result of filtering with a GLPF with \( D_0 = 80 \). Note reduction in skin fine lines in the magnified sections of (b) and (c).
Sharpening Frequency Domain Filter

Ideal highpass filter

\[ H_{hp}(u, v) = H_{lp}(u, v) \]

Butterworth highpass filter

\[
H(u, v) = \begin{cases} 
0 & \text{if } D(u, v) \leq D_0 \\
1 & \text{if } D(u, v) > D_0 
\end{cases}
\]

\[
H(u, v) = \frac{1}{1 + \left(\frac{D_0}{D(u, v)}\right)^n}
\]

Gaussian highpass filter

\[
H(u, v) = 1 - e^{-D^2(u,v)/2D_0^2}
\]

FIGURE 4.22 Top row: Perspective plot, image representation, and cross section of a typical ideal highpass filter. Middle and bottom rows: The same sequence for typical Butterworth and Gaussian highpass filters.
**FIGURE 4.23** Spatial representations of typical (a) ideal, (b) Butterworth, and (c) Gaussian frequency domain highpass filters, and corresponding gray-level profiles.
Ideal Highpass Filters

\[ H(u, v) = \begin{cases} 
0 & \text{if } D(u, v) \leq D_0 \\
1 & \text{if } D(u, v) > D_0 
\end{cases} \]

**Figure 4.24** Results of ideal highpass filtering the image in Fig. 4.11(a) with \( D_0 = 15, 30, \) and 80, respectively. Problems with ringing are quite evident in (a) and (b).
Butterworth Highpass Filters

\[ H(u, v) = \frac{1}{1 + \left( \frac{D_0}{D(u, v)} \right)^{2n}} \]

**FIGURE 4.25** Results of highpass filtering the image in Fig. 4.11(a) using a BHPF of order 2 with \( D_0 = 15, 30, \) and \( 80, \) respectively. These results are much smoother than those obtained with an ILPF.
Gaussian Highpass Filters

\[ H(u, v) = 1 - e^{-D^2(u,v)/2D_0^2} \]

**FIGURE 4.26** Results of highpass filtering the image of Fig. 4.11(a) using a GHPF of order 2 with \( D_0 = 15, 30, \) and \( 80, \) respectively. Compare with Figs. 4.24 and 4.25.
The Laplacian in the Frequency Domain

- The Laplacian filter

\[ H(u, v) = -(u^2 + v^2) \]

- Shift the center:

\[ H(u, v) = -\left[ (u - \frac{M}{2})^2 + (v - \frac{N}{2})^2 \right] \]
\[ g(x, y) = f(x, y) - \nabla^2 f(x, y) \]

where

\[ \nabla^2 f(x, y) : \text{the Laplacian-filtered image in the spatial domain} \]
Some Additional Properties of the 2D Fourier Transform

- Periodicity, symmetry, and back-to-back properties

**Figure 4.34**
(a) Fourier spectrum showing back-to-back half periods in the interval \([0, M - 1]\).
(b) Shifted spectrum showing a full period in the same interval.
(c) Fourier spectrum of an image, showing the same back-to-back properties as (a), but in two dimensions.
(d) Centered Fourier spectrum.
Some Additional Properties of the 2D Fourier Transform

- Separability

**Figure 4.35**
Computation of the 2-D Fourier transform as a series of 1-D transforms.
## Summary of Some Important Properties of the 2-D Fourier Transform

<table>
<thead>
<tr>
<th>Property</th>
<th>Expression(s)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Fourier transform</td>
<td>[ F(u, v) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) e^{-j2\pi(ux/M+vy/N)} ]</td>
</tr>
<tr>
<td>Inverse Fourier transform</td>
<td>[ f(x, y) = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F(u, v) e^{j2\pi(ux/M+vy/N)} ]</td>
</tr>
<tr>
<td>Polar representation</td>
<td>[ F(u, v) =</td>
</tr>
<tr>
<td>Spectrum</td>
<td>[</td>
</tr>
<tr>
<td>Phase angle</td>
<td>[ \phi(u, v) = \tan^{-1}\left( \frac{I(u, v)}{R(u, v)} \right) ]</td>
</tr>
<tr>
<td>Power spectrum</td>
<td>[ P(u, v) =</td>
</tr>
<tr>
<td>Average value</td>
<td>[ \bar{f}(x, y) = F(0, 0) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) ]</td>
</tr>
</tbody>
</table>
| Translation               | \[ f(x, y) e^{j2\pi(ux_0/M+vy_0/N)} \iff F(u-u_0, v-v_0) \]
|                           | \[ f(x-x_0, y-y_0) \iff F(u, v) e^{-j2\pi(ux_0/M+vy_0/N)} \]
|                           | When \( x_0 = u_0 = M/2 \) and \( y_0 = v_0 = N/2 \), then \[ f(x, y)(-1)^{x+y} \iff F(u-M/2, v-N/2) \]
|                           | \[ f(x-M/2, y-N/2) \iff F(u, v)(-1)^{u+v} \]
## Summary of Some Important Properties of the 2-D Fourier Transform

<table>
<thead>
<tr>
<th>Property</th>
<th>Formula</th>
</tr>
</thead>
<tbody>
<tr>
<td>Conjugate symmetry</td>
<td>$F(u, v) = F^*(-u, -v)$, $</td>
</tr>
<tr>
<td>Differentiation</td>
<td>$rac{\partial^n f(x, y)}{\partial x^n} \leftrightarrow (ju)^n F(u, v)$, $(-jx)^n f(x, y) \leftrightarrow \frac{\partial^n F(u, v)}{\partial u^n}$</td>
</tr>
<tr>
<td>Laplacian</td>
<td>$\nabla^2 f(x, y) \leftrightarrow -(u^2 + v^2) F(u, v)$</td>
</tr>
<tr>
<td>Distributivity</td>
<td>$\Re[f_1(x, y) + f_2(x, y)] = \Re[f_1(x, y)] + \Re[f_2(x, y)]$, $\Im[f_1(x, y) \cdot f_2(x, y)] \neq \Im[f_1(x, y)] \cdot \Im[f_2(x, y)]$</td>
</tr>
<tr>
<td>Scaling</td>
<td>$af(x, y) \leftrightarrow aF(u, v)$, $f(ax, by) \leftrightarrow \frac{1}{</td>
</tr>
<tr>
<td>Rotation</td>
<td>$x = r \cos \theta$, $y = r \sin \theta$, $u = \omega \cos \varphi$, $v = \omega \sin \varphi$, $f(r, \theta + \theta_0) \leftrightarrow F(\omega, \varphi + \theta_0)$</td>
</tr>
<tr>
<td>Periodicity</td>
<td>$F(u, v) = F(u + M, v) = F(u, v + N) = F(u + M, v + N)$, $f(x, y) = f(x + M, y) = f(x, y + N) = f(x + M, y + N)$</td>
</tr>
<tr>
<td>Separability</td>
<td>See Eqs. (4.6-14) and (4.6-15). Separability implies that we can compute the 2-D transform of an image by first computing 1-D transforms along each row of the image, and then computing a 1-D transform along each column of this intermediate result. The reverse, columns and then rows, yields the same result.</td>
</tr>
</tbody>
</table>
### Summary of Some Important Properties of the 2-D Fourier Transform

<table>
<thead>
<tr>
<th>Property</th>
<th>Expression(s)</th>
</tr>
</thead>
</table>
| Computation of the inverse Fourier transform using a forward transform algorithm | \[
\frac{1}{MN} f^*(x, y) = \frac{1}{MN} \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F^*(u, v) e^{-j2\pi(ux/M + vy/N)}
\]
This equation indicates that inputting the function $F^*(u, v)$ into an algorithm designed to compute the forward transform (right side of the preceding equation) yields $f^*(x, y)/MN$. Taking the complex conjugate and multiplying this result by $MN$ gives the desired inverse. |
| Convolution†                          | \[
f(x, y) * h(x, y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f(m, n) h(x - m, y - n)
\]
| Correlation‡                          | \[
f(x, y) \circ h(x, y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f^*(m, n) h(x + m, y + n)
\]
| Convolution theorem†                  | \[
f(x, y) * h(x, y) \iff F(u, v) H(u, v);
\]
| Correlation theorem†                  | \[
f(x, y) \circ h(x, y) \iff F^*(u, v) H(u, v);
\]

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Summary of Some Important Properties of the 2-D Fourier Transform

<table>
<thead>
<tr>
<th>Function</th>
<th>Fourier Transform</th>
</tr>
</thead>
<tbody>
<tr>
<td>Impulse</td>
<td>$\delta(x, y) \leftrightarrow 1$</td>
</tr>
<tr>
<td>Gaussian</td>
<td>$A \sqrt{2\pi} \sigma e^{-2\pi^2 \sigma^2(x^2+y^2)} \leftrightarrow Ae^{-(u^2+v^2)/2\sigma^2}$</td>
</tr>
<tr>
<td>Rectangle</td>
<td>$\text{rect}[a, b] \leftrightarrow ab \frac{\sin(\pi u a)}{(\pi u a)} \frac{\sin(\pi v b)}{(\pi v b)} e^{-j\pi(ua+vb)}$</td>
</tr>
<tr>
<td>Cosine</td>
<td>$\cos(2\pi u_0 x + 2\pi v_0 y) \leftrightarrow \frac{1}{2} [\delta(u + u_0, v + v_0) + \delta(u - u_0, v - v_0)]$</td>
</tr>
<tr>
<td>Sine</td>
<td>$\sin(2\pi u_0 x + 2\pi v_0 y) \leftrightarrow j\frac{1}{2} [\delta(u + u_0, v + v_0) - \delta(u - u_0, v - v_0)]$</td>
</tr>
</tbody>
</table>

‡ Assumes that functions have been extended by zero padding.